

Introduction to calculating the digitally controlled PID filter

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http://www.optoelec-engineering.com/index_en.html

1. Introduction

We supply the application sample code of DSP / TMS320F28335 as to the digitally controlled PID filter via our “***DSP F28335 Basic Control Platform C-Programming Code Sets***”.

We provide another detail documents about feedback control. Please check the following document.

“Introduction to feed back control.pdf”

The sample application of the digitally controlled PID filter is the incomplete integral type such as Fig- 1. We get the transfer function of Fig- 1 with replacing these variables such as eq.- 1 for the Fig-2 and eq.-17 of “***Introduction to feed back control.pdf***”.

R3→R1, R2→open, C2→open, R1→R2, C1→C1, Vref→0

eq.- 1

Additionally, we consider the open-loop gain A_v of OP-Amp. Thus, we make the equation as to transfer function for Fig- 1.

$$i_1 = i_3 = \frac{V_{in} - V_c}{R_1} = \frac{V_c - V_{out}}{R_2 + \frac{1}{j\omega C_1}}$$

eq.- 2

$$V_{out} = A_v \times (0 - V_c)$$

eq.- 3

Substitute eq.- 2 to eq.- 3. Here, $s=j\omega=j \times 2\pi f$.

$$\begin{aligned} \frac{V_{in} + \frac{V_{out}}{A_v}}{R_1} &= \frac{-\frac{V_{out}}{A_v} - V_{out}}{R_2 + \frac{1}{sC_1}} \\ -\left(R_2 + \frac{1}{sC_1}\right) \cdot \left(\frac{V_{in}}{R_1} + \frac{V_{out}}{A_v R_1}\right) &= \left(\frac{1}{A_v} + 1\right) \cdot V_{out} \\ -\frac{R_2}{R_1} \cdot \left(1 + \frac{1}{sC_1 R_2}\right) \cdot \left(V_{in} + \frac{V_{out}}{A_v}\right) &= \left(\frac{1}{A_v} + 1\right) \cdot V_{out} \\ -\frac{R_2}{R_1} \cdot \left(1 + \frac{1}{sC_1 R_2}\right) \cdot V_{in} &= \left(\frac{1}{A_v} + 1\right) \cdot V_{out} + \frac{R_2}{R_1} \cdot \left(1 + \frac{1}{sC_1 R_2}\right) \cdot \frac{V_{out}}{A_v} \end{aligned}$$

Therefore, we get the transfer function of PID filter Fig- 1.

$$\begin{aligned}
G_{PID}(s) &= \frac{V_{out}}{V_{in}} = \frac{-\frac{R_2}{R_1} \cdot \left(1 + \frac{1}{sC_1R_2}\right)}{\frac{R_2}{R_1} \cdot \left(1 + \frac{1}{sC_1R_2}\right) \cdot \frac{1}{A_v} + \frac{1}{A_v} + 1} \\
&= \frac{-\frac{R_2}{R_1} \cdot sC_1R_2 - \frac{R_2}{R_1}}{\frac{R_2}{R_1} \cdot sC_1R_2 \cdot \frac{1}{A_v} + \frac{R_2}{R_1} \cdot \frac{1}{A_v} + \left(\frac{1}{A_v} + 1\right) \cdot sC_1R_2} \\
&= \frac{-\frac{C_1R_2}{R_1} \cdot s - \frac{1}{R_1}}{\frac{C_1R_2}{R_1} \cdot \frac{1}{A_v} \cdot s + \frac{1}{R_1} \cdot \frac{1}{A_v} + \left(\frac{1}{A_v} + 1\right) \cdot sC_1} \\
&= \frac{-C_1R_2 \cdot s - 1}{C_1R_2 \cdot \frac{1}{A_v} \cdot s + \frac{1}{A_v} + \left(\frac{1}{A_v} + 1\right) \cdot sC_1R_1} \\
&= -\frac{sC_1R_2 + 1}{s \cdot \left((C_1R_2 + C_1R_1) \frac{1}{A_v} + C_1R_1 \right) + \frac{1}{A_v}} \\
&= -\frac{sC_1R_2 + 1}{s \cdot C_1 \left((R_2 + R_1) \frac{1}{A_v} + R_1 \right) + \frac{1}{A_v}} \\
&= \frac{-A_v \cdot (s \cdot C_1R_2 + 1)}{s \cdot C_1 (R_2 + (1 + A_v) \cdot R_1) + 1}
\end{aligned}$$

eq.- 4

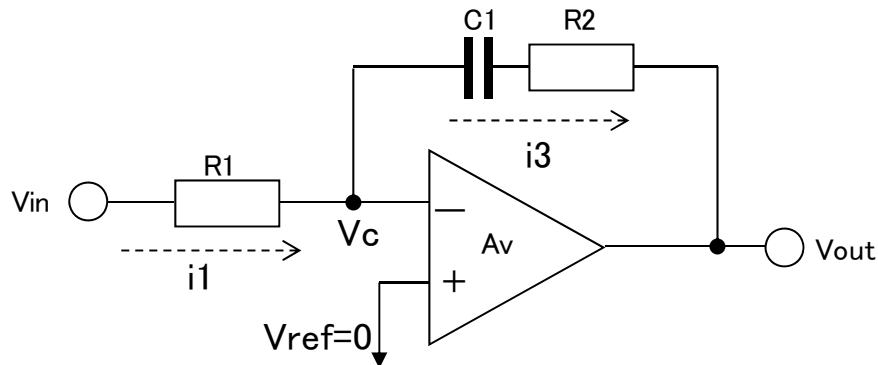


Fig- 1 PID filter of the incomplete integral type provided in
“DSP F28335 Basic Control Platform C-Programming Code Sets”

Here, we get a zero-point and a pole-point from eq.- 4.

$$\text{Zero-point} \quad f_{c2} = \frac{1}{2\pi C_1 R_2}$$

eq.- 5

$$\text{Pole-point} \quad f_{c1} = \frac{1}{2\pi C_1 (R_2 + (1 + A_v) R_1)} \approx \frac{1}{2\pi A_v C_1 R_1}$$

eq.- 6

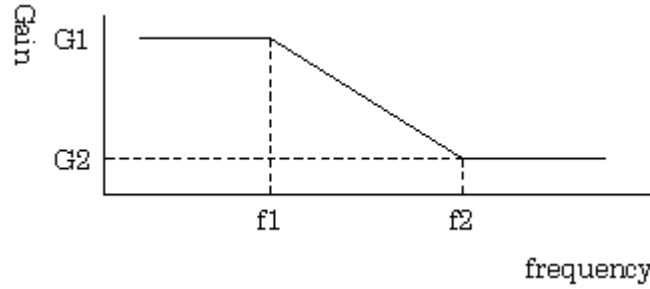


Fig- 2 Transfer characteristics of PID filter Fig- 1

We calculate the amplitude term and the phase term of eq.- 4.

First, we replace eq.- 4 by eq.- 7.

$$P = -A_v C_1 R_2 \quad Q = C_1 (R_2 + (1 + A_v) R_1)$$

eq.- 7

Then, we get eq.- 8.

$$G_{PID}(s) = \frac{V_{out}}{V_{in}} = \frac{-A_v \cdot (s \cdot C_1 R_2 + 1)}{s \cdot C_1 (R_2 + (1 + A_v) \cdot R_1) + 1} = \frac{s \cdot P - A_v}{s \cdot Q + 1}$$

eq.- 8

And more, we expand eq.- 8.

$$\begin{aligned} G_{PID}(s) &= \frac{V_{out}}{V_{in}} = \frac{s \cdot P - A_v}{s \cdot Q + 1} = \frac{(s \cdot P - A_v) \cdot (s \cdot Q - 1)}{(s \cdot Q + 1) \cdot (s \cdot Q - 1)} = \frac{s^2 \cdot PQ - s \cdot (A_v Q + P) + A_v}{s^2 \cdot Q^2 - 1} \\ &= \frac{-\omega^2 \cdot PQ - s \cdot (A_v Q + P) + A_v}{-\omega^2 \cdot Q^2 - 1} = \frac{\omega^2 \cdot PQ - A_v}{\omega^2 \cdot Q^2 + 1} + j \times \omega \cdot \frac{A_v Q + P}{\omega^2 \cdot Q^2 + 1} \end{aligned}$$

eq.- 9

The amplitude component (Gain) is

$$\text{Gain} \left[\frac{V}{V} \right] = \sqrt{\left(\frac{\omega^2 \cdot PQ - A_v}{\omega^2 \cdot Q^2 + 1} \right)^2 + \left(\omega \cdot \frac{A_v Q + P}{\omega^2 \cdot Q^2 + 1} \right)^2} = \frac{\sqrt{(\omega^2 \cdot PQ - A_v)^2 + \omega^2 \cdot (A_v Q + P)^2}}{\omega^2 \cdot Q^2 + 1}$$

eq.- 10

The phase component is

$$Phase[rad] = \tan^{-1} \left(\frac{\omega \cdot \frac{A_v Q + P}{\omega^2 \cdot Q^2 + 1}}{\frac{\omega^2 \cdot PQ - A_v}{\omega^2 \cdot Q^2 + 1}} \right) = \tan^{-1} \left(\frac{\omega \cdot (A_v Q + P)}{(\omega^2 \cdot PQ - A_v)} \right)$$

eq.- 11

2. Principle of Digitally controlled PID filter

We start to discretize the transfer function eq.- 9 of analog PID filter by Bilinear Transform eq.- 12.

$$\text{Bilinear Transform (s} \rightarrow z^{-1}\text{)} \quad s = \frac{2}{T_s} \cdot \frac{1 - z^{-1}}{1 + z^{-1}}$$

eq.- 12

Here each parameter is,

Analog complex angular frequency $s = j\omega = j \times 2\pi f$

T_s ; a time of sampling period

z^{-1} ; delay-operator eq.- 13

$$z^{-1} = e^{-j\Omega} = \cos(\Omega) - j \cdot \sin(\Omega)$$

eq.- 13

Ω is digital angular frequency.

$$\Omega = 2 \tan^{-1} \left(\frac{T_s \cdot \omega}{2} \right)$$

eq.- 14

Thus, analog complex angular frequency s and analog angular frequency ω is

$$s = j\omega = j \cdot \frac{2}{T_s} \times \tan\left(\frac{\Omega}{2}\right) \quad \omega = \frac{2}{T_s} \cdot \tan\left(\frac{\Omega}{2}\right)$$

eq.- 15

Substitute eq.- 12 into eq.- 8.

$$G_{PID}(s) = \frac{s \cdot P - A_v}{s \cdot Q + 1} = \frac{\frac{2}{T_s} \cdot \frac{1 - z^{-1}}{1 + z^{-1}} \cdot P - A_v}{\frac{2}{T_s} \cdot \frac{1 - z^{-1}}{1 + z^{-1}} \cdot Q + 1} = \frac{\frac{2}{T_s} P \cdot (1 - z^{-1}) - A_v \cdot (1 + z^{-1})}{\frac{2}{T_s} Q \cdot (1 - z^{-1}) + (1 + z^{-1})}$$

eq.- 16

Here, replace new parameter of to eq.- 16

$$\xi = \frac{2P}{T_s} \quad \chi = \frac{2Q}{T_s}$$

eq.- 17

Thus

$$G_{PID}(z) = \frac{OUT(z)}{IN(z)} = \frac{\frac{2}{T_s} P \cdot (1 - z^{-1}) - A_v \cdot (1 + z^{-1})}{\frac{2}{T_s} Q \cdot (1 - z^{-1}) + (1 + z^{-1})} = \frac{\xi \cdot (1 - z^{-1}) - A_v \cdot (1 + z^{-1})}{\chi \cdot (1 - z^{-1}) + (1 + z^{-1})}$$

eq.- 18

Expand the eq.- 18.

$$OUT(z) \cdot \{\chi \cdot (1 - z^{-1}) + (1 + z^{-1})\} = IN(z) \cdot \{\xi \cdot (1 - z^{-1}) - A_v \cdot (1 + z^{-1})\}$$

$$(\chi + 1) \cdot OUT(z) + (-\chi + 1) \cdot z^{-1} OUT(z) = (\xi - A_v) \cdot IN(z) - (\xi + A_v) \cdot z^{-1} IN(z)$$

eq.- 19

Therefore,

$$OUT(z) = \frac{\xi - A_v}{\chi + 1} \cdot IN(z) - \frac{\xi + A_v}{\chi + 1} \cdot z^{-1} IN(z) + \frac{\chi - 1}{\chi + 1} \cdot z^{-1} OUT(z)$$

$$= a \times IN(z) + b \times z^{-1} IN(z) + c \times z^{-1} OUT(z)$$

eq.- 20

Here, a, b, c constants are

$$a = \frac{\xi - A_v}{\chi + 1} \quad b = -\frac{\xi + A_v}{\chi + 1} \quad c = \frac{\chi - 1}{\chi + 1}$$

eq.- 21

z^{-1} is delay-operator. The $z^{-1} \cdot IN(z)$ is previous sampling input data with one-time sampling delay, the $z^{-1} \cdot OUT(z)$ is previous sampling output data with one-time sampling delay.

The circuit diagram of eq.- 20 are wrote on Fig- 3.

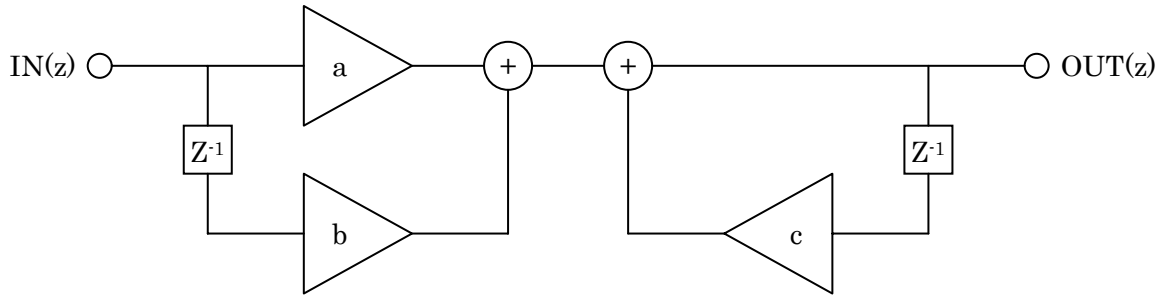


Fig- 3 Diagram of digitally controlled PID filter

We transform the equation eq.- 18 to a format of digital angular frequency Ω . Substitute eq.- 13 to eq.- 18. We get

$$G(z) = \frac{\{\xi(1 - \cos(\Omega)) - A_v(1 + \cos(\Omega))\} + j \times \sin(\Omega) \times (\xi + A_v)}{\{\chi(1 - \cos(\Omega)) + (1 + \cos(\Omega))\} + j \times \sin(\Omega) \times (\chi - 1)} = \frac{D + j \cdot E}{F + j \cdot G} = \frac{DF + EG}{F^2 + G^2} + j \times \frac{EF - DG}{F^2 + G^2}$$

eq.- 22

Here,

$$D = \xi(1 - \cos(\Omega)) - A_v(1 + \cos(\Omega))$$

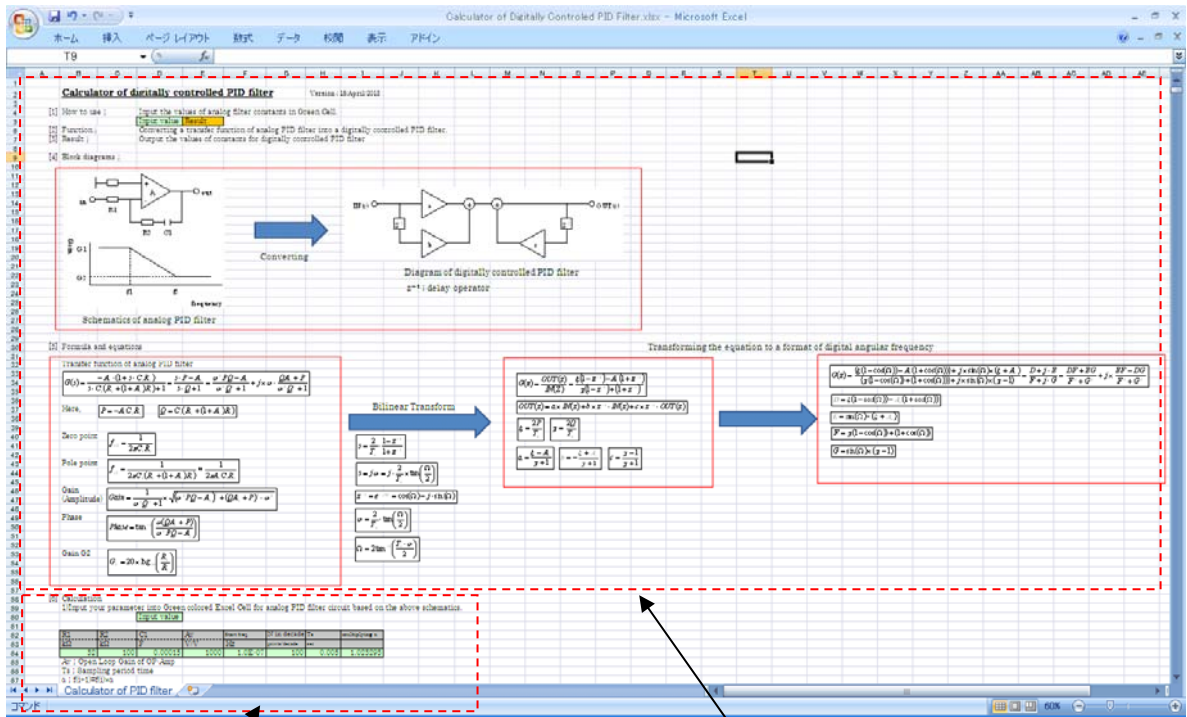
$$E = \sin(\Omega) \times (\xi + A_v)$$

$$F = \chi(1 - \cos(\Omega)) + (1 + \cos(\Omega))$$

$$G = \sin(\Omega) \times (\chi - 1)$$

eq.- 23

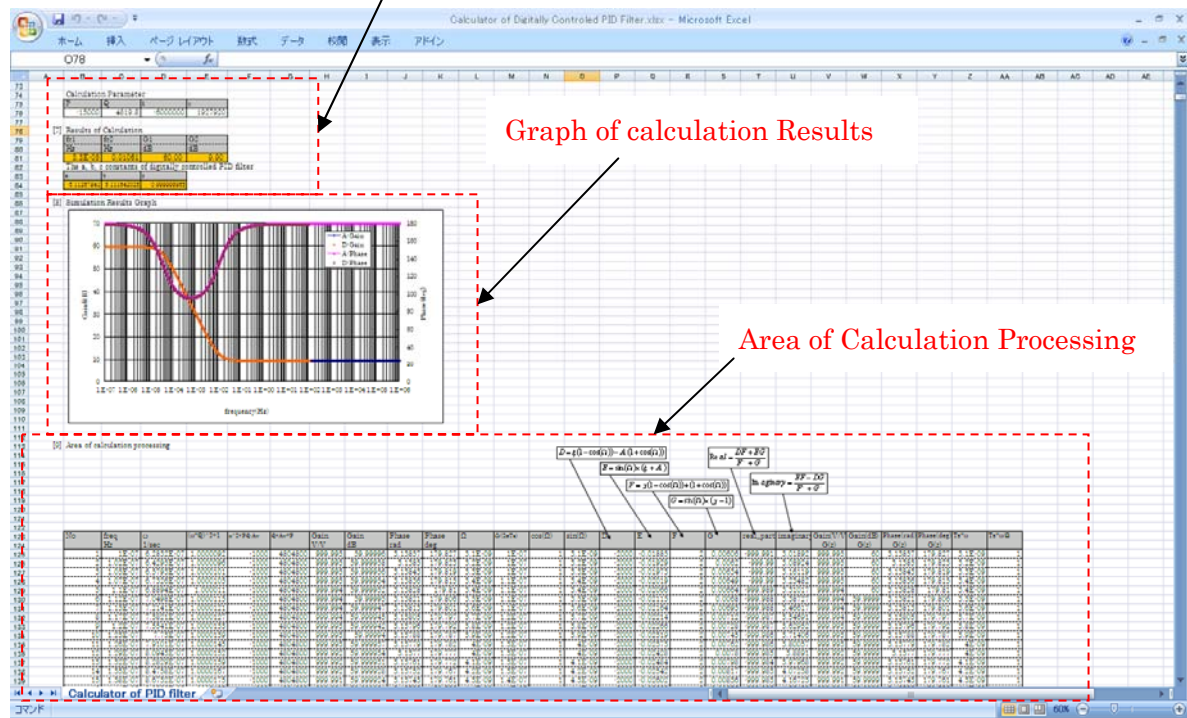
3. How to use our “Calculator of digitally controlled PID filter”. (Excel files)



Area of Input

Area of calculation Results

Area of Explanations



Graph of calculation Results

Area of Calculation Processing

Fig. 4 Whole view of Excel Calculator

[1] Input your values of analog filter constants in Green Cell of excel calculator.

[2] User can get the results of calculator.

[3] Please read the comments of the excel calculator.

4. Our websales site

We sell the “*Calculator of digitally controlled PID filter*” and its document “*Introduction to calculating the digitally controlled PID filter*”. Please visit our websales site.

http://www.optoelec-engineering.com/websale/websale_en.html